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INTERFACE UNIT FOR SERIAL-TO-PARALLEL CONVERSION AND/OR
PARALLEL-TO-SERIAL CONVERSION

59 BACKGROUND OF THE INVENTION

1. Field of the Invention

The present invention relates to a communication interface circuit; and more particularly, to a communication interface circuit which permits selection of the amount of data converted from serial-to-parallel and from parallel-to-serial.

2. Description of Related Art.

Fig. 1 is a schematic block diagram of a codec interface circuit 1 connected to a conventional serial communication interface circuit 2. As shown, the codec interface circuit 1 and the conventional serial communication interface circuit 2 are disposed on different chips which are then connected together. In Fig. 1, a clock divider 20 receives a master clock signal CLK and an initial count setting SET. The clock divider 20 divides the master clock CLK into a plurality of low frequency clocks. In Fig. 1, the symbol phi represents the master clock CLK, but of a different phase.

A clock source selecting unit 30 receives the plurality of clocks output from the clock divider 20, and based on a selection value stored in a register 32 thereof

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selects and outputs one of the received clocks. By selecting the plurality of received clocks based on a register value, flexibility exists in setting the system speed. Simply by changing the register value, a designer can alter the system speed.

A frame generating unit 34 receives the selected clock SCLK and converts this clock to an even lower frequency clock SYNC. The selected clock SCLK also clocks a transmit unit 36 and a receive unit 38. As shown, the transmit unit 36 receives a write signal and a read signal. The transmit unit 36 is in parallel communication with a data bus, and has a serial output TXout. The amount of serial data transferred between the data bus and the transmit unit 36 is fixed. Conventionally, the width of the parallel data is fixed at either 8 bits or 16 bits. In the conventional serial communication interface circuit 2 of Fig. 1, the width is shown as 16 bits.

The receive unit 38 also receives the read signal, receives parallel communication from the data bus, and has a serial input RXin. Like the transmit unit 36, the width of the parallel communication received by the receive unit 38 is fixed, and is fixed to the same width as the communication between the data bus and the transmit unit 36.

The operation of the transmit unit 36 and the receive unit 38 will be described in detail below with respect to Figs. 2 and 3. Figs. 2 and 3 illustrate circuit diagrams of the transmit unit 36 and the receive unit 38, respectively.

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A shown in Fig. 1, the transmit unit 36 includes a first transmit shift register TX1 and a second transmit shift register TX2. Both the first and second transmit shifter registers TX1 and TX2 have the same 8 bit storage capacity, receive the selected clock signal SCLK, receive the read signal, and receive the write signal. The first transmit shift register TX1 is connected to the eight most significant bits of the 16 bit wide data bus, and the second transmit shift register TX2 is connected to the eight least significant bits of the 16 bit wide data bus. The serial input of the first transmit shift register TX1 is connected to ground, and the serial input of the second transmit shift register TX2 is connected to the serial output of the first transmit shift register TX1. The serial output of the second transmit shift register TX2 serves as the output of the transmit unit 36. Typically, both the first and second transmit shift registers TX1 and TX2 are composed of eight 1 bit shift registers connected in series.

When a logic high write signal is received, the first and second transmit shift registers TX1 and TX2 input the parallel data on the data bus. Then, with each pulse of the selected clock SCLK, the first and second transmit shift registers TX1 and TX2 serially shift the data stored therein from their serial inputs to their serial outputs. Accordingly, as the eight bits of parallel data are serially shifted out from the first transmit shift register TX1 to the second transmit shift register TX2, logic level low data is shifted into the first transmit shift register TX1 because the serial input thereof is connected to ground. Meanwhile, as the data stored in the

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second transmit shift register TX2 is shifted out, the serial data from the first transmit shift register TX1 is shifted in. As the shifting of data out of the second transmit shift register TX1 continues, the serial data from the first transmit shift register TX1 is eventually shifted out of the second transmit shift register TX2. After sixteen pulses of the selected clock SCLK, the parallel data originally input by the first and second transmit shift registers TX1 and TX2 is output as serial data.

When the read signal is logic level low, no parallel input or serial output of data takes place. When the first and second transmit shift registers TX1 and TX2 receive a logic high read signal, the data stored therein is output in parallel to the data bus.

Referring to Fig. 3, the receive unit 38 includes a first receive shift register RX1 and a second receive shift register RX2. Both the first and second receive shift registers RX1 and RX2 have the same 8 bit storage capacity, receive the selected clock signal SCLK, and receive the read signal. The first receive shift register RX1 is connected to the eight most significant bits of the 16 bit wide data bus, and the second receive shift register RX2 is connected to the eight least significant bits of the 16 bit wide data bus. The serial input of the second receive shift register RX2 is connected to the serial output of the first receive shift register RX1. The serial input of the first receive shift register RX1 serves as the serial input for the receive unit 38. Both the write enable inputs of the first and

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second receive shift registers RX1 and RX2 are disabled by being connected to ground. Typically, both the first and second receive shift registers RX1 and RX2 are composed of eight 1 bit shift registers connected in series.

When a logic high read signal is received, the first and second receive shift registers RX1 and RX2 shift data from their serial inputs to their serial outputs in accordance with the selected clock signal SCLK. Because each of the first and second receive shift registers RX1 and RX2 is eight bits wide, it takes eight pulses of the selected clock signal SCLK for data to transfer across one of the first and second receive shift registers RX1 and RX2. After 16 pulses of the selected clock signal SCLK both the first and second receive shift registers RX1 and RX2 are filled with new serial data. Then, the first and second receive shift registers RX1 and RX2 transfer the data stored therein in parallel to the data bus.

The eight bit serial communication interface has the same structure as the 16 bit serial communication interface described above except that the transfer and receive units in the eight bit serial communication interface include a single transmit shift register and a single receive shift register, respectively.

Depending on the design at issue, an operator must select between using an 8 or 16 bit serial communication interface. It is desirable, however, for an operator to be able to use a single interface, and then selectively set the interface to either an 8 bit or 16 bit operating mode. Furthermore, while illustrated as two chips, it would also be preferable in terms of improving integration and improving

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efficiency to place the codec interface circuit and the serial communication interface circuit on a single chip.

SUMMARY OF THE INVENTION

One object of the present invention is to overcome the disadvantages and drawbacks of conventional serial communication interfaces.

Another object of the present invention is to provide a serial communication interface which allows an operator to select between operation in different bit length modes.

A further operation of the present invention is to provide a serial communication interface which also provides the function of a codec interface on the same chip.

These and other objects are achieved by providing a data conversion interface, comprising: a clock signal generator generating a clock signal in response to a mode signal, said mode signal indicating operation in one of at least a first and second data length transfer mode; a serial-to-parallel converter receiving the clock signal, the mode signal and serial data, and converting the serial data into parallel data having a data length as set forth in the mode signal.

These and other objects are further achieved by providing a data conversion interface, comprising: a clock signal generator generating a clock signal in response to a mode signal, said mode signal indicating operation in one of at least

a first and second data length transfer mode; and a parallel-to-serial converter receiving the clock signal, the mode signal and parallel data, and converting the parallel data into serial data having a data length as set forth in the mode signal.

5 DA BRIEF DESCRIPTION OF THE DRAWINGS

The accompanying drawings, which are included to provide a further understanding of the invention and are incorporated in and constitute a part of this specification, illustrate embodiments of the invention and together with the description serve to explain the principles of the invention.

In the drawings:

Fig. 1 is a schematic block diagram of a conventional serial communication interface circuit;

Fig. 2 is a circuit diagram of the transmit unit of Fig. 1:

Fig. 3 is a circuit diagram of the receive unit of Fig. 1:

Fig. 4 is a schematic block diagram of an embodiment of a serial communication interface circuit according to the present invention;

Fig. 5 illustrates the codec interface unit in Fig. 4;

Fig. 6 illustrates the transmit unit of Fig. 4; and

Fig. 7 illustrates the receive unit of Fig. 4.

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DE DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

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Fig. 4 illustrates a schematic block diagram of an embodiment of a serial communication interface according to the present invention. Where the serial communication interface of this embodiment includes the same components as in the conventional serial communication interface of Fig. 1, the same reference numerals have been used to designate those components. Additionally, because of their prior description, the operation of these components will not be described in detail.

As shown, the serial communication interface of Fig. 4 includes a codec interface unit 10 which receives the master clock CLK and generates the low frequency clock SYNC and an intermediate frequency clock CDCLK. Fig. 5 illustrates the codec interface unit 10 in detail. As shown, the codec interface unit 10 includes a mod 3 counter 12 generating a mod 3 count value in accordance with the master clock signal CLK. A T-type flip flop (T-ff) 14 is triggered to change a logic state of it output on a rising edge of the output from the mod 3 counter 12. Accordingly, the mod 3 counter 12 and the T-FF 14 operate to produce an intermediate frequency clock CDCLK having a frequency less than the hamaster clock CLK. The frame generating unit 34 then generates the low frequency clock SYNC from the intermediate frequency clock CDCLK.

Returning to Fig. 4, a clock divider 20 receives the master clock signal CLK and an initial count setting SET. The clock divider 20 divides the master clock CLK into a plurality of low frequency clocks. In Fig. 4, the symbol phi represents

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the master clock CLK as before

A clock source selecting unit 30 receives the plurality of clocks output from the clock divider 20 including the intermediate frequency clock CDCLK output from the codec interface unit 10 and the master clock CLK (note the master clock and the clock labeled phi from the clock divider 20 are the same clock, but have a different phase), and based on a selection value stored in a register 32 thereof selects and outputs one of the received clocks. By selecting the plurality of received clocks based on a register value, flexibility exists in setting the system speed. Simply by changing the register value, a designer can alter the system speed.

A ternary/tetrad counter 40 receives the selected clock SCLK, the set signal SET and a mode signal. The mode signal indicates whether the serial communication interface should operate in an 8 bit mode or a 16 bit mode. When the mode signal indicates operation in the 8 bit mode, the ternary/tetrad counter 40, based on the initial value established by the set signal SET, counts in mod 3 in accordance with the selected clock SCLK. When the mode signal indicates operation in the 16 bit mode, the ternary/tetrad counter 40, based on the initial value established by the set signal SET, counts in mod 4 in accordance with the selected clock SCLK. Accordingly, one skilled in the art will appreciate that the ternary/tetrad counter 40 generates 8 pulses in a predetermined period of time when acting as a mod 3 counter and generates 16 pulses in the same

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predetermined period of time when acting as a mod 4 counter. The count value output from the ternary/tetrad counter 40 serves as an operation clock SIOCLK for a transmit unit 52 and a receive unit 54.

As further shown in Fig. 4, the transmit unit 52 receives a write signal and a read signal. The transmit unit 52 is in parallel communication with a data bus, and has a serial output TXout. The amount of serial data transferred between the data bus and the transmit unit 52 is not fixed, but is selectable between either 8 bit or 16 bit widths.

The receive unit 54 also receives the read signal, receives parallel dath to communication from the data bus, and has a serial input RXin. Like the transmit unit 52, the width of the parallel communication is not fixed, but is selectable and set at the same width as the communication between the data bus and the transmit unit 52.

The operation of the transmit unit 52 and the receive unit 54 will be described in detail below with respect to Figs. 6 and 7. Figs. 6 and 7 illustrate circuit diagrams of the transmit unit 52 and the receive unit 54, respectively.

A shown in Fig. 6, the transmit unit 52 includes a first transmit shift register TX1 and a second transmit shift register TX2. Both the first and second transmit shift registers TX1 and TX2 have the same 8 bit storage capacity, receive the operation clock SIOCLK, receive the read signal, and receive the write signal. The first transmit shift register TX1 is connected to the eight most significant bits of

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the 16 bit wide data bus, and the second transmit shift register TX2 is connected to the eight least significant bits of the 16 bit wide data bus. The serial input of the first transmit shift register TX1 is connected to ground, and the serial input of the second transmit shift register TX2 is connected to the serial output of the first transmit shift register TX1 via a first enable unit 60. The serial output of the second transmit shift register TX2 serves as the output of the transmit shift register 36.

The first enable unit 60 includes a first AND gate 61 receiving the serial output of the first transmit shift register TX1 and the mode signal. A first inverter 63 also receives the mode signal, and a second AND gate 62 receives the output of the first inverter 63 and a ground voltage. A first OR gate 64 receives the output of the first and second AND gates 61 and 62, and the output of the first OR gate 64 is connected to the serial input of the second transmit unit TX2.

When a logic high write signal is received, the first and second transmit shift registers TX1 and TX2 input the parallel data en the data bus. Then, with each pulse of the selected clock, the first and second transmit shift registers TX1 and TX2 serially shift the data stored therein from their serial inputs to their serial outputs. Accordingly, the first AND gate 61 receives the eight bits of parallel data serially shifted out from the first transmit shift register TX1.

Because the second AND gate 62 always receives the logic level low ground signal, its output will always be logic level low. Consequently, the output of the

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OR gate 64 depends entirely on the output of the first AND gate 61. The mode signal is logic level high when indicating the 16 bit mode of operation, and thus, the output of the second AND gate 61 and the first OR gate 64 is determined by the serial output of the first transmit shift register TX1. Stated another way, in the 16 bit operating mode, the first enable unit 60 transfers the serial data output from the first transmit shift register TX1 to the serial input of the second transmit shift register TX2.

The mode signal, however, is logic low when indicating the 8 bit operating mode. Consequently, the output of the first AND gate 61 will always be logic level low. With the first OR gate 64 receiving logic low inputs from both the first and second AND gates 61 and 62, the output of the first OR gate 64 is continually logic low. Accordingly, in the 8 bit operating mode, the second transmit shift register TX2 shifts in a logic low data stream. A logic level low data stream is shifted into the first transmit shift register TX1, regardless of the operating mode, because the serial input thereof is connected to ground.

From the forgoing description, its is readily apparent that in the 8 bit operating mode, only the parallel data input by the first transmit shift register TX2 is output as serial data because the first enable unit 60 prevents the serial data output from the first transmit shift register TX1 from reaching the serial input of the second transmit shift register TX2. Also, because the ternary/tetrad counter 40 outputs a mod 3 count as the operation clock SIOCLK, during the

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predetermined period of time between inputting parallel data, the operation clock SIOCLK has 8 pulses; enough to shift the parallel data out of the second transmit shift register TX2. However, in the 16 bit mode, the ternary tetrad counter 40 outputs a mod 4 count as the operation clock SIOCLK. During the same predetermined period of time, the operation clock SIOCLK has 16 pulses.

Because in the 16 bit mode the serial output of the first transmit shift register TX1 is transferred by the first enable unit 60 to the second transmit shift register TX2, the parallel data input by both the first and second transmit shift registers TX1 and TX2 is output during the predetermined period of time.

When the first and second transmit shift register TX1 and TX2 receive a logic high read signal, the data stored therein in is output in parallel to the data bus, and no operation takes place when both the write and read signals are logic low.

Referring to Fig. 3, the receive unit 54 includes a first receive shift register RX1 and a second receive shift register RX2. Both the first and second receive shift registers RX1 and RX2 have the same 8 bit storage capacity, receive the operation clock SIOCLK, and receive the read signal. The first receive shift register RX1 is connected to the eight most significant bits of the 16 bit wide data bus, and the second receive shift register RX2 is connected the eight least significant bits of the 16 bit wide data bus. The serial input of the second receiver shift register RX2 is selectively connected to the serial input or the serial output

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of the first receive shift register RX1 via a second enable unit 80. The serial input of the first receive shift register RX1 serves as the serial input for the receive unit 54. Both the write enable inputs of the first and second receive shift registers RX1 and RX2 are disabled by being connected to ground.

The second enable unit 80 includes a third AND gate 81 receiving the serial output of the first receive shift register RX1 and the mode signal. A second inverter 83 also receives the mode signal, and a fourth AND gate 82 receives the output of the second inverter 83 and the serial data input to the first receive shift register RX1. A second OR gate 84 receives the outputs of the third and fourth AND gates 81 and 82.

When indicating the 16 bit operating mode, the mode signal is logic level high. The output from the second inverter 83 is logic level low, and as a result, the output of the fourth AND gate 82 is logic level low regardless of the state of the serial data input to the first receive shift register RX1. Therefore, the output of the second OR gate 84 is determined entirely by the output of the third AND gate 81. Because the mode signal is logic level high, the output of the third AND gate 81 is determined by the serial data output from the first receive shift register RX1. Stated another way, in the 16 bit operating mode, the second enable unit 80 transfers the serial data output from the first receive shift register RX1 to the serial input of the second receive shift register RX2.

When indicating the 8 bit operating mode, the mode signal is logic level

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low, and the output from the third AND gate 81 is logic level low regardless of the state of the data output from the first receive shift register RX1. Accordingly, the output from the second OR gate 84 is determined entirely by the output from the fourth AND gate 82. The output of the second inverter 83 is logic high, and thus, the output of the fourth AND gate 82 is determined entirely by the serial data input to the first receive shift register RX1. Stated another way, in the 8 bit operating mode, the second enable unit 80 transfers the serial data input to the first receive shift register RX1 to the serial input of the second receive shift register RX2.

When a logic high read signal is received, the first and second receive shift registers RX1 and RX2 shift data from their serial inputs to their serial outputs in accordance with the selected clock signal. Because each of the first and second receive shift registers RX1 and RX2 is eight bits wide, it take eight pulses of the operation clock signal SIOCLK for data to transfer across one of the first and second receive shift registers RX1 and RX2. In the 8 bit operating mode, after 8 pulses of the operation clock SIOCLK the first and second receive shift registers RX1 and RX2 are filled with the same new serial data. As discussed above, the eight pulses of the operation clock SIOCLK are received during a predetermined period of time. At the end of the predetermined period of time, the data stored in the first and second shift registers RX1 and RX2 is transferred in parallel to the data bus. Because only the lower 8 bits of the data bus are being used in the 8 bit

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operating mode, the transfer of data to the upper 8 bits of the data bus has no effect.

In the 16 bit operating mode, the first and second receive shift registers RX1 and RX2 receive 16 pulses of the operation clock SIOCLK during the predetermined period of time. As a result, after 16 pulses of the operation clock signal SIOCLK, both the first and second receive shift registers RX1 and RX2 are filled with new serial data. But since the second enable unit 80 transferred the serial output of the first receive shift register RX1 to the serial input of the second receive shift register RX2, the serial data filling each register is different. Then, at the end of the predetermined period of time, the first and second receive shift registers RX1 and RX2 transfer the data stored therein in parallel to the data bus.

When the read signal is low, no operation takes place, and changes in the write signal have no effect on the first and second receive shift registers RX1 and RX2.

As discussed above in detail, the serial communication interface according to the present invention can operate in either an 8 bit serial-to-parallel conversion mode or a 16 bit serial-to-parallel conversion mode, and can operate in either an 8 bit parallel-to-serial conversion mode or a 16 bit parallel-to-serial conversion mode. As one skilled in the art will appreciate, by increasing the number of bits in the mode signal, increasing the width of the data bus and adding receive and

transmit shift registers, the number of operating modes can be increased.

Furthermore, while the serial communication interface has been described as connected to a 16 bit wide data bus, the serial communication interface may also be connected to an 8 bit wide data bus.

It will be apparent to those skilled in the art that various modifications and variations can be made to the present invention without departing from the spirit or scope of the invention. Thus, it is intended that the present invention cover the modifications and variations of this invention provided they come within the scope of the appended claims and their equivalents.